

H. Tromp
 Laboratory of Electromagnetism and Acoustics
 University of Ghent, Ghent, Belgium

ABSTRACT

Exact formulas are derived for the upper and lower limits of the amplitude of the input reflection coefficient of an arbitrary two-port, in the presence of mismatched source and load and/or indefinite reference planes. The derivation is based on certain properties of the bilinear transformation.

Introduction

It was shown by Bandler, Liu and Tromp¹ that, in the realistic worst-case tolerance analysis of microwave two-ports, the effect of a mismatched source and load should not be neglected, as compared with the effect of physical tolerances and model uncertainties. Explicit formulas were derived for the extrema of the modulus of the input reflection coefficient of a lossless two-port, referred to real normalisation impedances. In this paper, we intend to generalise these formulas. We will consider the situation depicted in fig. 1. The S-matrix of the two-port is referred to Z'_1 and Z'_2 . Source and load impedances Z_S and Z_L are represented by their reflection coefficients, ρ_S and ρ_L , w.r.t. Z_1 and Z_2 respectively. Z_i and Z'_i may be complex. Let

$$\rho_S = |\rho_S| e^{j\phi_S}, \quad \rho_L = |\rho_L| e^{j\phi_L} \quad (1)$$

We assume ϕ_S , ϕ_L arbitrary, and $|\rho_S|$, $|\rho_L|$ either given, or limited by

$$0 \leq |\rho_S| \leq |\rho_S|^+, \quad 0 \leq |\rho_L| \leq |\rho_L|^+ \quad (2)$$

This corresponds to the realistic situation, where only a (maximum) VSWR is specified for source and load. We are interested in the input reflection coefficient,

$$\rho_{in} = \frac{Z_{in} - Z_S^*}{Z_{in} + Z_S^*} \quad (3)$$

and we shall derive expressions for

$$|\rho_{in}|_M = \max_{\phi_S, \phi_L} |\rho_{in}|, \quad |\rho_{in}|_m = \min_{\phi_S, \phi_L} |\rho_{in}|, \quad (4)$$

$$\text{or } |\rho_{in}|^+ = \max_{|\rho_S|, |\rho_L|, \phi_S, \phi_L} |\rho_{in}|, \quad (5)$$

$$|\rho_{in}|^- = \min_{|\rho_S|, |\rho_L|, \phi_S, \phi_L} |\rho_{in}|$$

Another situation which shall be dealt with is shown in fig. 2, where the lengths of the connecting lines at input and output (given by phase-angles ϕ_1 and ϕ_2) are arbitrary. We shall consider only real Z_i , Z'_i in this case.

Again, expressions for the limits (4)-(5) will be derived, but now with the extrema taken over

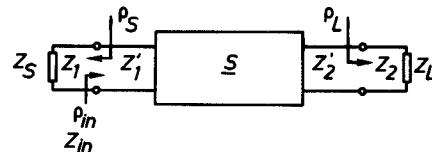


Fig.1. Two-port with mismatched source and load

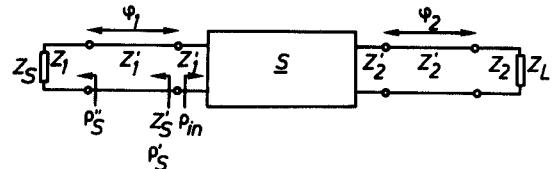


Fig.2. Two-port with arbitrary reference planes

ϕ_1 and ϕ_2 also.

The expressions to be derived are useful for worst-case analysis and fit into the general formulation of the tolerance problem, as given by Tromp². They also give the possibility, when performing a tolerance optimization (see e.g. ^{1,3-5}) to compromise between the tolerances within the network (i.e. its cost) and the quality of source and load. To a certain extent they can be used to design subnetworks of a large network separately. Finally, they offer an alternative for the study of the stability of a two-port under various conditions of source and load, as well as various positions of input and output reference planes.

Bilinear transformations

Consider the transformation (w, z, a, b, c, d) complex)

$$w = \frac{az+b}{cz+d} \quad (6)$$

This well-known bilinear transformation was, among others, studied by Deschamps^{6,7} and is known to transform circles into circles. Without loss of generality, we can consider

$$w = Re^{j\phi}, \quad 0 \leq \phi \leq 2\pi.$$

Then

$$z = z_o + r e^{j\theta}, \quad 0 \leq \theta \leq 2\pi \quad (7)$$

with

$$z_o = \frac{R^2 c^* d - a^* b}{|a|^2 - R^2 |c|^2}, \quad r = \frac{R |ad - bc|}{||a|^2 - R^2 |c|^2} \quad (8)$$

*Part of this work was supported by a grant of the NFWO (Belgian National Research Fund)

Then

$$|z_M| = \max_{\phi} |z| = |z_o| + r \quad (9)$$

$$|z_m| = \min_{\phi} |z| = |z_o| - r \quad (10)$$

One can prove that $|z_M|$ and $|z_m|$, as functions of R , behave as follows : $|z_M|$ increases to ∞ for $R \leq \frac{|a|}{|c|}$ and decreases for $R > \frac{|a|}{|c|}$, while $|z_m|$ decreases to 0 for $R \leq \frac{|b|}{|d|}$ and increases for $R > \frac{|b|}{|d|}$.

If we let $R^- \leq R \leq R^+$, then

$$|z|^+ = \max_{\phi, R} |z| = \begin{cases} |z_M(R^+)| & \text{if } R^+ < \frac{|a|}{|c|} \\ \infty & \text{if } R^- \leq \frac{|a|}{|c|} \leq R^+ \\ |z_M(R^-)| & \text{if } R^- > \frac{|a|}{|c|} \end{cases} \quad (11)$$

$$|z|^- = \min_{\phi, R} |z| = \begin{cases} |z_m(R^+)| & \text{if } R^+ < \frac{|b|}{|d|} \\ 0 & \text{if } R^- \leq \frac{|b|}{|d|} \leq R^+ \\ |z_m(R^-)| & \text{if } R^- > \frac{|b|}{|d|} \end{cases} \quad (12)$$

Two special cases of (9) and (10) are of interest

1) If $\arg a + \arg d = \arg b + \arg c + \pi$, then

$$|z_M| = \left| \frac{|b|+R|d|}{|a|-R|c|} \right|, \quad |z_m| = \left| \frac{|b|-R|d|}{|a|+R|c|} \right| \quad (13)$$

2) If $\arg a + \arg d = \arg b + \arg c$, then the value of $|z_M|$ and $|z_m|$ is given by Table 1, where

	$\left \frac{b}{a} \right < \left \frac{d}{c} \right $	$\left \frac{b}{a} \right > \left \frac{d}{c} \right $
$R < R_o$	z_B	z_A
$R > R_o$	z_A	z_B
z_M	z_B	z_A
z_m	z_A	z_B

Table 1

z_M and z_m when $\arg a + \arg d = \arg b + \arg c$

$$|z_A| = \left| \frac{|b|-R|d|}{|a|-R|c|} \right|, \quad |z_B| = \left| \frac{|b|+R|d|}{|a|+R|c|} \right| \quad (14)$$

and

$$R_o = \sqrt{\left| \frac{ab}{cd} \right|} \quad (15)$$

We consider also the transformation

$$z = \frac{w^* d^* - b^*}{a - w c} \quad (16)$$

which does not transform circles into circles, but gives the same $|z|$ as (6). This means that all results, (9) to (15) also apply for (16), which we therefore call the pseudo-bilinear transformation.

Effect of mismatches

In fig. 1, we have

$$|\rho_{in}| = \left| \frac{\rho - \rho_S^*}{1 - \rho \rho_S} \right| \quad (17)$$

with

$$\rho = \frac{z_{in} - z_1}{z_{in} + z_1^*} \quad (18)$$

We can now use (16), to find the extrema of $|\rho_{in}|$, for all ρ_S , and either for a given ρ , or for $|\rho|$ between upper and lower limits. ρ depends only on ρ_L , which means that $|\rho|$ can be forced to its extreme values, independently of ρ_S .

$$\text{Indeed, } \rho_L = \frac{a \rho + b}{c \rho + d} \quad (19)$$

with

$$a = (1 + S_{11}' \Delta_1) (1 + S_{22}' \Delta_2) - S_{12}' S_{21}' \Delta_1 \Delta_2 \quad (20)$$

$$b = -(\Delta_1^* + S_{11}') (1 + S_{22}' \Delta_2) + S_{12}' S_{21}' \Delta_2 \quad (21)$$

$$c = (1 + S_{11}' \Delta_1) (\Delta_2^* + S_{22}') - S_{12}' S_{21}' \Delta_1 \quad (22)$$

$$d = -(\Delta_1^* + S_{11}') (\Delta_2^* + S_{22}') + S_{12}' S_{21}' \quad (23)$$

$$S_{11}' = \gamma_1^* S_{11}, \quad S_{22}' = \gamma_2^* S_{22}, \quad S_{12}' S_{21}' = \gamma_1^* \gamma_2^* S_{12} S_{21} \quad (24)$$

$$\gamma_i = \frac{z_i^* + z_i}{z_i^* + z_i}, \quad \Delta_i = \frac{z_i^* - z_i}{z_i^* + z_i}, \quad i = 1, 2 \quad (25)$$

(19) is a bilinear transformation and (9) to (12) give the limits of $|\rho|$, for all ρ_L . Finally, we find

$$|\rho_{in}|_M = \begin{cases} K^+ (|\rho_M|, |\rho_S|) & \text{if } |\rho_M| < 1 \leq \frac{1}{|\rho_S|} \\ K^- (|\rho_M|, |\rho_S|) & \text{if } |\rho_M| < \frac{1}{|\rho_S|} \leq 1 \\ \text{or if } 1 \leq |\rho_M| < \frac{1}{|\rho_S|} \\ K^+ (|\rho_m|, |\rho_S|) & \text{if } \frac{1}{|\rho_S|} \leq 1 < |\rho_m| \\ K^- (|\rho_m|, |\rho_S|) & \text{if } \frac{1}{|\rho_S|} < |\rho_m| \leq 1 \\ \text{or if } 1 \leq \frac{1}{|\rho_S|} < |\rho_m| \\ \infty & \text{if } |\rho_m| \leq \frac{1}{|\rho_S|} \leq |\rho_M| \end{cases} \quad (26)$$

and similar expressions for $|\rho_{in}|_m$, with $|\rho_M|$, $|\rho_m|$ according to (9) and (10).

Also

$$|\rho_{in}|^+ = \begin{cases} K^+ (|\rho|^+, |\rho_S|^+) & \text{if } |\rho_L|^+ < \left| \frac{a}{c} \right| \\ \text{and } |\rho|^+ < 1 \leq \frac{1}{|\rho_S|^+} \\ K^- (|\rho|^+, |\rho_S|^+) & \text{if } |\rho_L|^+ < \left| \frac{a}{c} \right| \text{ and} \\ \text{either } |\rho|^+ < \frac{1}{|\rho_S|^+} \leq 1 & \text{or } 1 \leq |\rho|^+ < \frac{1}{|\rho_S|^+} \\ \infty & \text{if } |\rho_L|^+ > \left| \frac{a}{c} \right| \\ \text{or } |\rho|^+ > \frac{1}{|\rho_S|^+} \end{cases} \quad (27)$$

and similar expressions for $|\rho_{in}|^-$, with $|\rho|^-$

and $|\rho|^-$ according to (11) and (12) (with $R^- = 0$, $R^+ = |\rho_L|^{+}$). We defined

$$K^+(x_1, x_2) = \left| \frac{x_1 + x_2}{1 + x_1 x_2} \right|, \quad K^-(x_1, x_2) = K^+(x_1, -x_2) \quad (28)$$

Undefined reference planes

Consider fig. 2, with Z_i and Z'_i real. Let ρ'_S be the reflection coefficient of Z'_S w.r.t Z'_1 , and ρ''_S this of Z_S w.r.t Z'_1 . Then

$$|\rho'_S| = |\rho''_S| \quad \text{and} \quad \rho_S = \frac{\rho''_S + \Delta_1}{1 + \rho''_S \Delta_1} \quad (29)$$

We can use (9) to (12) to find the extrema of $|\rho'_S|$ over all ρ_S . The same can be done at the load. A situation similar to that of the preceding paragraph arises and similar formulas are found.

Lossless two-port

If $Z'_i = Z_i$ (real) and if the network is lossless, special case 2) is found and the formulas are considerably simplified. If Z_S and Z_L are passive, the formulas derived in¹ are found.

Example

As an example, we consider the transistor HP 35821E (bias $I_C = 15\text{mA}$, $V_{CE} = 15\text{V}$). Fig. 3 gives the limits of $|\rho_{in}|$, $|\rho_{in}|_M$ and $|\rho_{in}|^-$ coincide, as well as $|\rho_{in}|_m$ and $|\rho_{in}|^-$. The two-port remains stable. Fig. 4 illustrates the effect of undefined reference planes. The two-port remains stable. In some cases however, the two-port becomes unstable when the reference planes are made arbitrary. The results were confirmed by a Monte Carlo analysis.

Conclusion

Exact, explicit formulas were derived for the limits of the input reflection coefficient of an arbitrary two-port, under various conditions of source and load. They can easily be implemented in a computer program.

References

- 1 J.W. Bandler, P.C. Liu and H. Tromp, "Integrated approach to microwave design", *IEEE Trans. Microwave Theory Techn.*, vol. MTT-24, Sept. 1976, pp. 584-591.
- 2 H. Tromp, "The generalized tolerance problem and worst case search", *Proc. Conf. on Computer Aided Design of Electronic, Microwave Circuits and Systems*, Hull, England, July 1977, pp. 72-77.
- 3 J.W. Bandler, P.C. Liu and H. Tromp, "A non-linear programming approach to optimal design centering, tolerancing and tuning", *IEEE Trans. Circuits and Systems*, vol. CAS-23, March 1976, pp. 155-165.
- 4 J.F. Pinel and K.A. Roberts, "Tolerance assignment in linear networks using nonlinear programming", *IEEE Trans. Circuit Theory*, vol. CT-19, Sept. 1972, pp. 475-479.
- 5 J.W. Bandler, P.C. Liu and J.H.K. Chen, "Worst case network tolerance optimization",

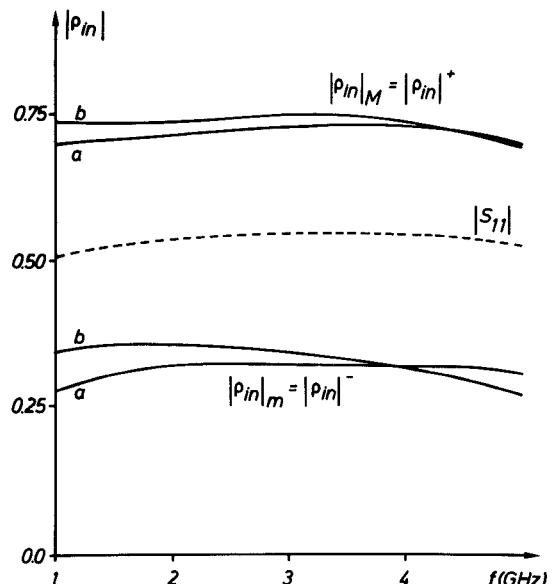


Fig.3. Input reflection coefficient of 35821E, with

$$|\rho_S|+ = |\rho_L|+ = 0.2$$

a) $Z_i = Z'_i = 50$
b) $Z_1 = 50 + 0.5j, Z_2 = 49 - 2j, Z'_1 = 45 + 5j, Z'_2 = 55 - 7j$

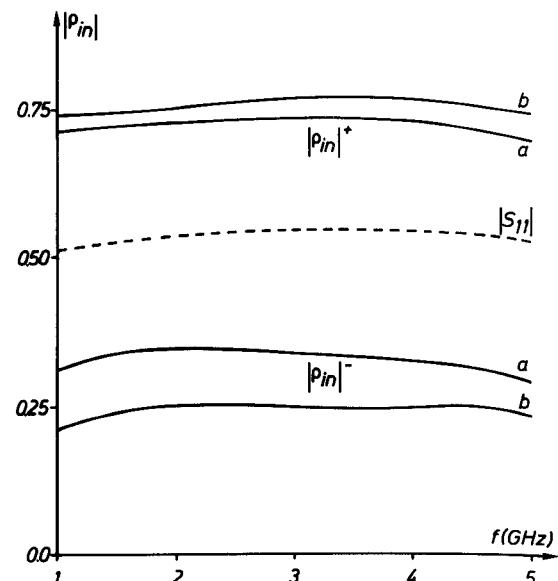


Fig.4. Input reflection coefficient of 35821E, with

$$|\rho_S|+ = |\rho_L|+ = 0.2, Z_1 = 50, Z'_1 = 45, Z_2 = 49, Z'_2 = 55$$

a) fixed reference planes
b) indefinite reference planes

IEEE Trans. Microwave Theory Techn., vol. MTT-23, Aug. 1975, pp. 630-641.

- 6 G.A. Deschamps, "Geometric viewpoints in the representation of waveguides and waveguide junctions", *Proc. Symp. on Modern Network Synthesis*, New York, Sept. 1952, pp. 277-295
- 7 G.A. Deschamps, "New chart for the solution of transmission-line and polarisation problems", *Trans. IRE Microwave Theory Techn.*, vol. 1, March 1953, pp. 5-13.